

General Description

The HM5301 is a 3CH power management IC for applications powered by one Li-Ion battery or a DC 5V adapter. It integrates three synchronous buck converters and can provide high efficiency output at light load and heavy load operation. The internal compensation architecture simplifies the application circuit design. Besides, the independent enable control makes the designer have the greatest flexibility to optimize timing for power sequencing purposes. The HM5301 is available in a 20 pin QFN 3x3 package.

Features

- ▶ 2.7V to 5.5V Input Voltage Range
- ▶ Three Buck Converters
 - Output Voltage Range: 0.6V to Vin
 - CH1, CH2, CH3 Continuous Load Current / Peak Current: 1A, 1A, 1A / 1.2A, 1.5A, 1.2A (3CH total output power consumption must be less than 6W)
 - Fixed 1.5MHz Switching Frequency
 - 100% Duty Cycle Low Dropout Operation
 - <1uA Shutdown Current
 - Independent Enable Control
 - Internal Compensation
 - Cycle-by-Cycle Current Limit
 - Short Circuit Protection
- ▶ Auto Recovery OTP Protection
- ▶ Input OVP Architecture
- ▶ Available in 20-pin 3mm x 3mm QFN Package

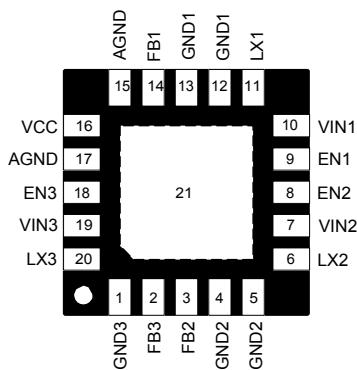
Applications

- ▶ Smart Phone
- ▶ IP Camera
- ▶ Digital Camera



Pin Configurations

(TOP VIEW)



QFN 3x3-20

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Pin Description

Pin Name	Function Description	Pin No.
GND3	Power ground pin of CH3.	1
FB3	Feedback input of CH3. Connect to output voltage with a resistor divider.	2
FB2	Feedback input of CH2. Connect to output voltage with a resistor divider.	3
GND2	Power ground pin of CH2.	4, 5
LX2	Internal MOSFET switching output of CH2. Connect LX2 pin with a low pass filter circuit to obtain a stable DC output voltage.	6
VIN2	Power input pin of CH2. Recommended to use a 10uF MLCC capacitor between VIN2 pin and GND2 pin.	7
EN2	CH2 turns on/turns off control input. Don't leave this pin floating.	8
EN1	CH1 turns on/turns off control input. Don't leave this pin floating.	9
VIN1	Power input pin of CH1. Recommended to use a 10uF MLCC capacitor between VIN1 pin and GND1 pin.	10
LX1	Internal MOSFET switching output of CH1. Connect LX1 pin with a low pass filter circuit to obtain a stable DC output voltage.	11
GND1	Power ground pin of CH1.	12, 13
FB1	Feedback input of CH1. Connect to output voltage with a resistor divider.	14
AGND	Analog ground pin.	15, 17
VCC	Input supply pin for internal control circuit.	16
EN3	CH3 turns on/turns off control input. Don't leave this pin floating.	18
VIN3	Power input pin of CH3. Recommended to use a 10uF MLCC capacitor between VIN3 pin and GND3 pin.	19
LX3	Internal MOSFET switching output of CH3. Connect SW3 pin with a low pass filter circuit to obtain a stable DC output voltage.	20
Exposed Pad	The Exposed Pad must be soldered to a large PCB copper plane and connected to GND for appropriate dissipation.	21

Function Block Diagram

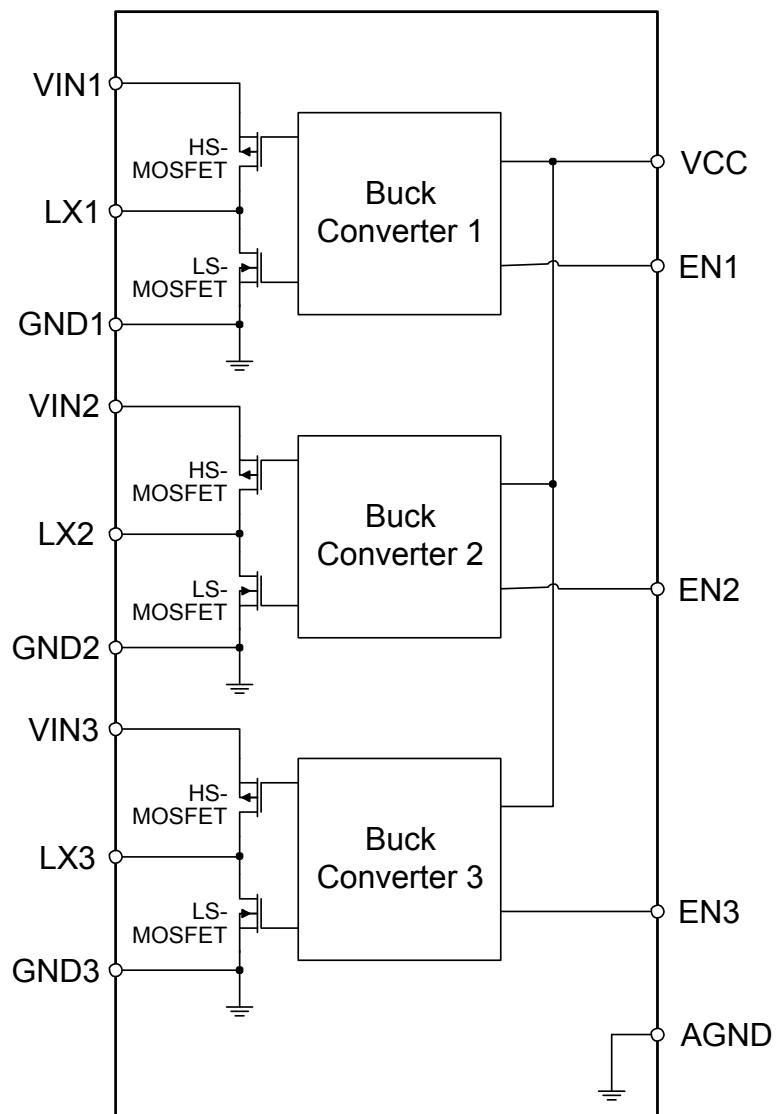


Figure 1. HM5301 internal function block diagram

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Absolute Maximum Ratings

Parameter	Value
Input Voltage (V_{VIN1} , V_{VIN2} , V_{VIN3} , V_{VCC})	-0.3V to +6.5V
SW Pin Voltage (V_{LX1} , V_{LX2} , V_{LX3})	-0.3V to $V_{VINX}+0.3V$
All Other Pins Voltage	-0.3V to +6.5V
Ambient Temperature operating Range (T_A)	-40°C to +85°C
Maximum Junction Temperature (T_{Jmax})	+150°C
Lead Temperature (Soldering, 10 sec)	+260°C
Storage Temperature Range (T_S)	-65°C to +150°C

Note (1): Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to "Absolute Maximum Ratings" conditions for extended periods may affect device reliability and lifetime.

Package Thermal Characteristics

Parameter	Value
QFN 3x3-20 Thermal Resistance (θ_{JC})	7.5°C/W
QFN 3x3-20 Thermal Resistance (θ_{JA})	67°C/W
QFN 3x3-20 Power Dissipation at $T_A=25^\circ C$ (P_{Dmax})	1.87W

Note (1): P_{Dmax} is calculated according to the formula: $P_{DMAX}=(T_{JMAX}-T_A)/\theta_{JA}$.

Recommended Operating Conditions

Parameter	Value
Input Voltage (V_{VIN1} , V_{VIN2} , V_{VIN3} , V_{VCC})	+2.7V to +5.5V
Junction Temperature Range (T_J)	-40°C to +125°C

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Electrical Characteristics

$V_{INX}=5V$, $V_{VCC}=5V$, $T_A=25^\circ C$, unless otherwise noted

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
Input Supply Voltage						
Input Voltage	V_{INX}		2.7		5.5	V
Control Circuit Input Voltage	V_{VCC}		2.7		5.5	V
Quiescent Current	$I_{Q\text{-total}}$	$V_{FB1} = V_{FB2} = V_{FB3} = 1V$		185	250	uA
Buck Converter 1, 2, 3						
Shutdown Supply Current	I_{SD}	$V_{EN} = 0V$	0.1	1		uA
UVLO Threshold	V_{UVLO}	V_{VIN} Rising	1.7	1.9	2.1	V
UVLO Hysteresis	$V_{UV\text{-HYST}}$			0.1		V
Output Load Current	I_{LOAD}			1	1.5	A
Reference Voltage	V_{REF}		0.588	0.6	0.612	V
Switching Frequency	F_{SW}	$I_{LOAD} = 100mA$	1	1.5	2	MHz
Short Frequency	$F_{SW\text{-SHORT}}$	$V_{OUT} = 0V$		330		KHz
CH1, CH3 PMOS Current Limit	$I_{LIM\text{-P}}$		1.5	2		A
CH2 PMOS Current Limit	$I_{LIM\text{-P}}$		1.7	2.2		A
PMOS On-Resistance	$R_{DS(ON)\text{-P}}$	$I_{LOAD} = 100mA$		120		mΩ
NMOS On-Resistance	$R_{DS(ON)\text{-N}}$	$I_{LOAD} = 100mA$		110		mΩ
Enable Pin Input Low Voltage	$V_{EN\text{-L}}$			0.4		V
Enable Pin Input High Voltage	$V_{EN\text{-H}}$		2			V
Maximum Duty Cycle	D_{MAX}		100			%
Input OVP Threshold Voltage	V_{OVP}	V_{IN} rising		6.3		V
Input OVP Threshold Hysteresis	$V_{OVP\text{-HYST}}$			150		mV
Thermal Shutdown						
Thermal Shutdown Threshold	T_{OTP}			165		°C
Thermal Shutdown Hysteresis	T_{HYST}			30		°C

Note (1): MOSFET on-resistance specifications are guaranteed by correlation to wafer level measurements.

(2): Thermal shutdown specifications are guaranteed by correlation to the design and characteristics analysis.

Application Circuit Diagram

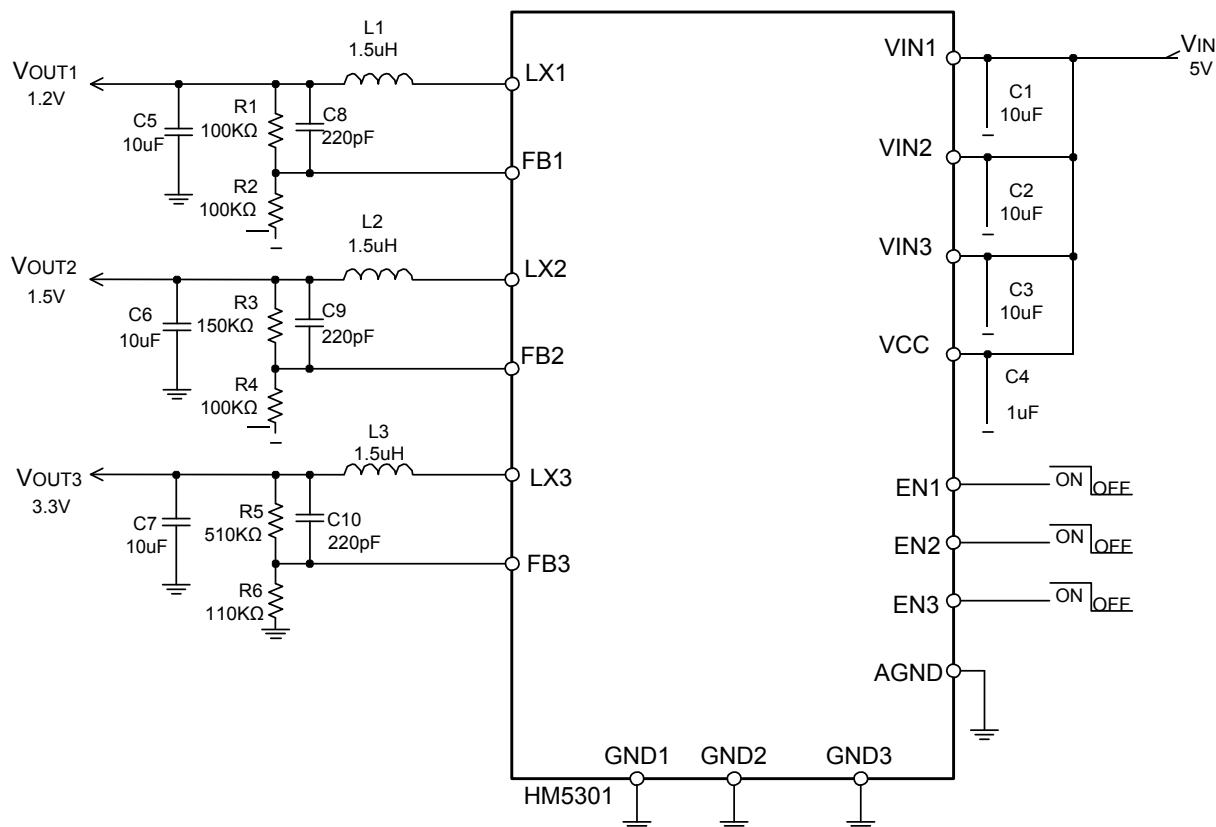


Figure 2. Typical application circuit diagram

Ordering Information

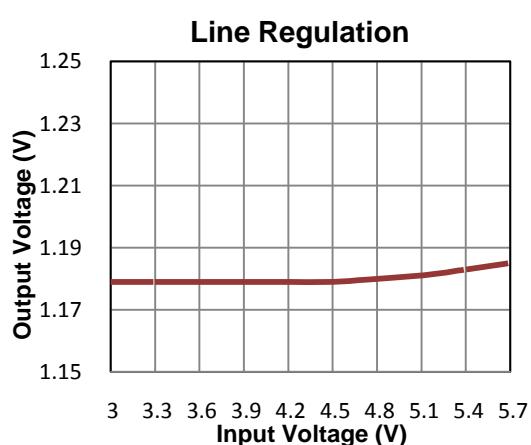
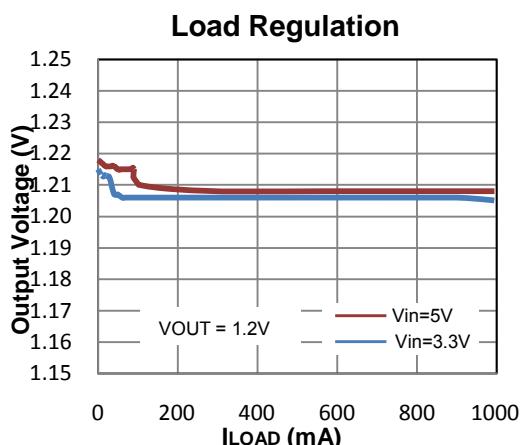
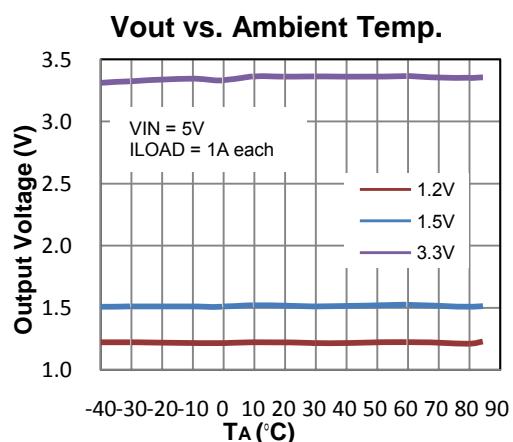
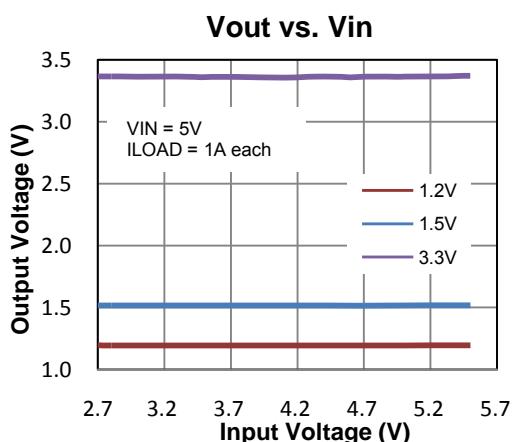
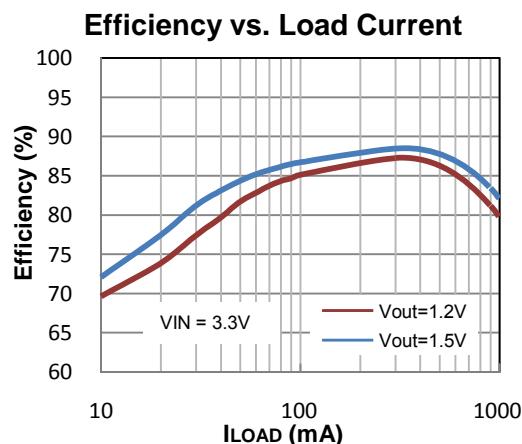
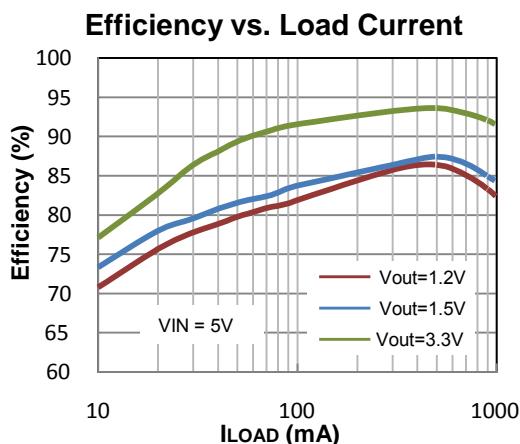
Part Number	Package Type	Packing Information
HM5301ÜÜ	QFN 3mm x 3mm-20	Tape & Reel / 3000

Note (1): "Q": Package type code.

(2): "R": Tape & Reel.

Typical Operating Characteristics

$V_{IN}=5V$, $V_{VCC}=5V$, $V_{OUT1}=3.3V$, $V_{OUT2}=1.5V$, $V_{OUT3}=1.2V$, $L1=1.5\mu H$, $L2=1.5\mu H$, $L3=1.5\mu H$, $T_A=25^\circ C$, unless otherwise noted

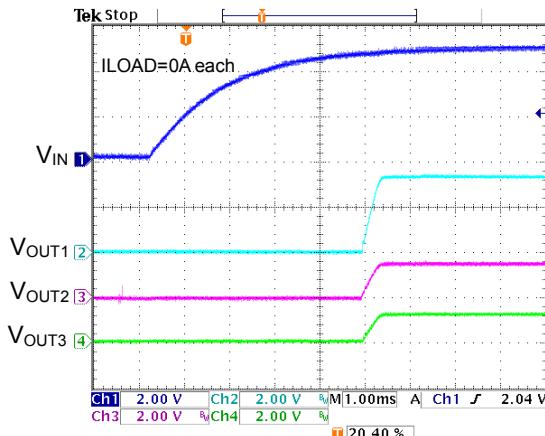


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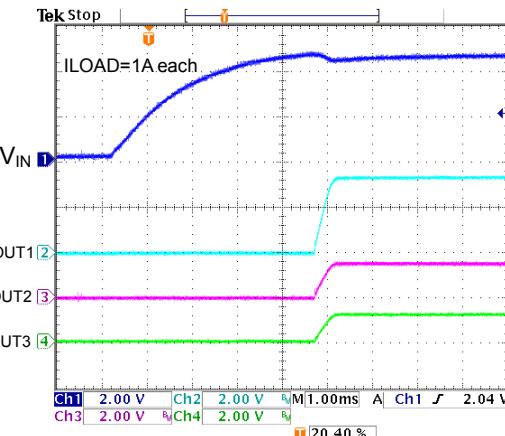
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Typical Operating Characteristics

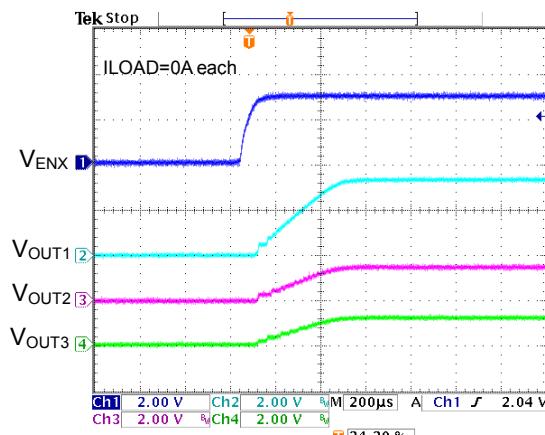
$V_{IN}=5V$, $V_{VCC}=5V$, $V_{OUT1}=3.3V$, $V_{OUT2}=1.5V$, $V_{OUT3}=1.2V$, $L1=1.5\mu H$, $L2=1.5\mu H$, $L3=1.5\mu H$, $T_A=25^\circ C$, unless otherwise noted



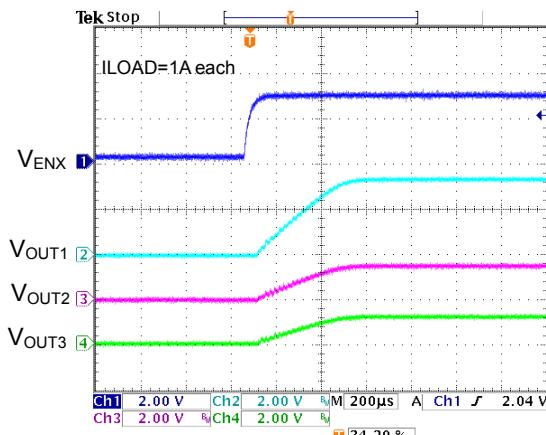
VIN Power On Waveform



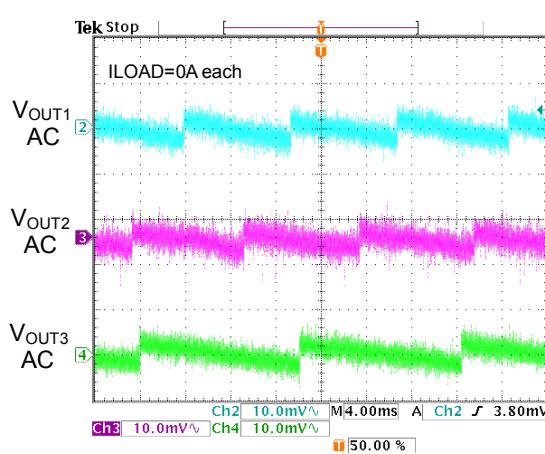
VIN Power On Waveform



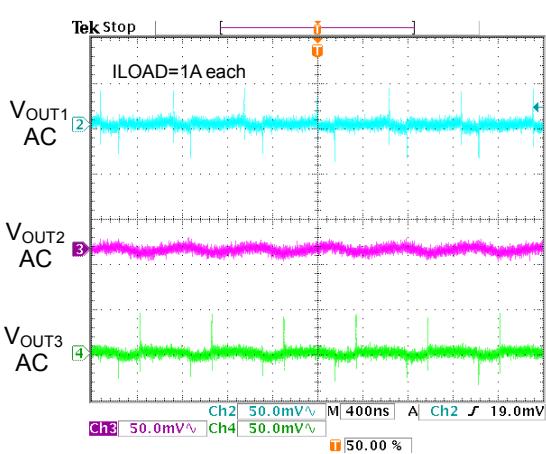
EN Power On Waveform



EN Power On Waveform



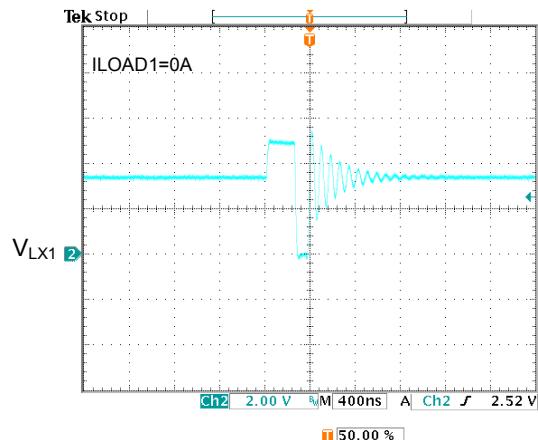
Output Ripple Waveform



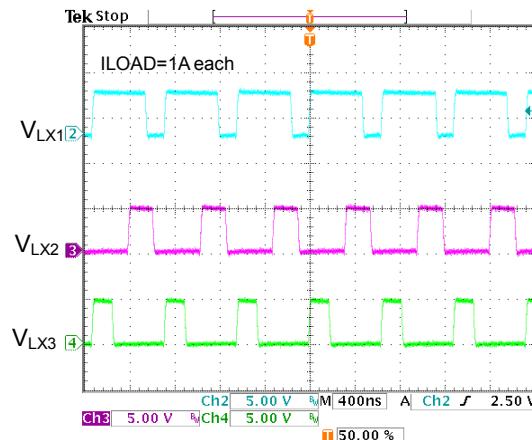
Output Ripple Waveform

Typical Operating Characteristics

$V_{IN}=5V$, $V_{VCC}=5V$, $V_{OUT1}=3.3V$, $V_{OUT2}=1.5V$, $V_{OUT3}=1.2V$, $L1=1.5\mu H$, $L2=1.5\mu H$, $L3=1.5\mu H$, $T_A=25^\circ C$, unless otherwise noted



V_{LX1} Switching Waveform



$V_{LX1,3}$ Switching Waveform

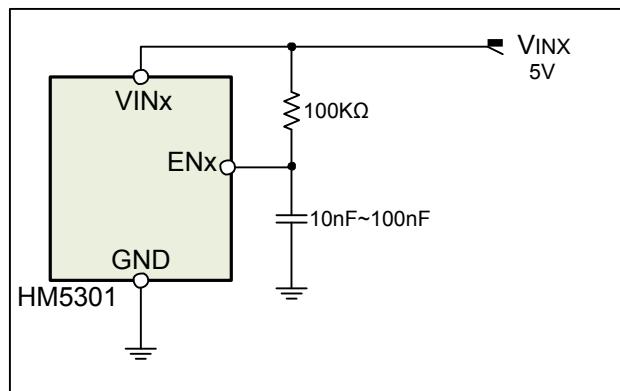
Functional Description

PFM/PWM Operation

Each of the buck regulators can be operated at PFM/PWM mode. If the output current is less than 150mA (typ.), the regulators automatically enters the PFM mode. The output voltages and output ripples at PFM mode are higher than the output voltages and output ripples at PWM mode. But at very light load, the PFM mode operation provides higher efficiency than PWM mode operation.

Enable Control

The HM5301 is a high efficiency Power Management IC which is designed for IPC applications. It incorporates three 1A synchronous buck regulators and can be controlled by individual EN pins. The start-up time for each channel can be programmed by using the circuit shown as below:



180° Phases Shifted Architecture

In order to reduce the input ripple current, the HM5301 applied 180° phases shifted architecture. Buck1 and Buck3 have the same phase and Buck2 is 180° out of phase. This architecture allows the system board has less ripple current, and thus can reduce EMI.

Over Current Protection

The HM5301 internal three regulators have their own cycle-by-cycle current limit circuits. When the inductor peak current exceeds the current limit threshold, the output voltage starts to drop until FB pin voltage is below the threshold, typically 30% below the reference. Once the threshold is triggered, the switching frequency is reduced to 330KHz (typ.).

Thermal Shutdown

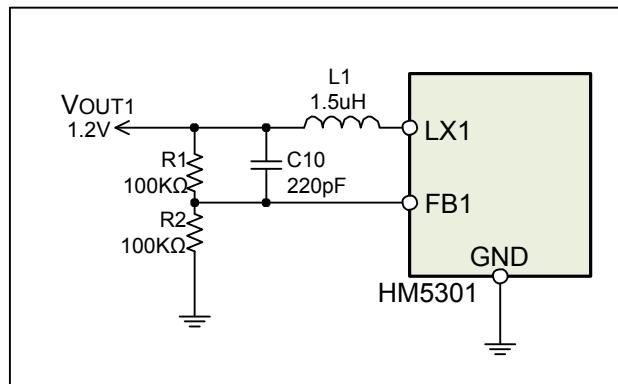
The HM5301 will automatically disabled if the die temperature is higher than the thermal shutdown threshold point. To avoid unstable operation, the hysteresis of thermal shutdown is about 30°C.

Application Information

Output Voltage Setting

Each of the regulators output voltage can be set via a resistor divider (ex. R1, R2). The output voltage is calculated by following equation:

$$V_{OUT1} = 0.6 \times \frac{R1}{R2} + 0.6 \text{ V}$$



The following table lists common output voltage and the corresponding R1, R2 resistance value for reference.

Output Voltage	R1 Resistance	R2 Resistance	Tolerance
3.3V	510KΩ	110KΩ	1%
1.8V	200KΩ	100KΩ	1%
1.5V	150KΩ	100KΩ	1%
1.2V	100KΩ	100KΩ	1%

Input / Output Capacitors Selection

The input capacitors are used to suppress the noise amplitude of the input voltage and provide a stable and clean DC input to the device. Because the ceramic capacitor has low ESR characteristic, so it is suitable for input capacitor use. It is recommended to use X5R or X7R MLCC capacitors in order to have better temperature performance and smaller capacitance tolerance. In order to suppress the output voltage ripple, the MLCC capacitor is also the best choice. The suggested part numbers of input / output capacitors are as follows:

Vendor	Part Number	Capacitance	Edc	Parameter	Size
TDK	C2012X5R1A106M	10uF	10V	X5R	0805
TDK	C3216X5R1A106M	10uF	10V	X5R	1206
TDK	C2012X5R1A226M	22uF	10V	X5R	0805
TDK	C3216X5R1A226M	22uF	10V	X5R	1206

Output Inductor Selection

The output inductor selection mainly depends on the amount of ripple current through the inductor ΔI_L . Large ΔI_L will cause larger output voltage ripple and loss, but the user can use a smaller inductor to save cost and space. On the contrary, the larger inductance can get smaller ΔI_L and

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thus the smaller output voltage ripple and loss. But it will increase the space and the cost. The inductor value can be calculated as:

$$L_{SW} = \frac{V_{PWR} - V_{OUT}}{I_{OUT}} \times \frac{V_{OUT}}{V_{PWR}}$$

For most applications, 1.0uH to 2.2uH inductors are suitable for HM5301.

Power Dissipation

The total output power dissipation of HM5301 should not exceed the maximum 6W range. The total output power dissipation can be calculated as:

$$P_D(\text{total}) = V_{OUT1} \times I_{OUT1} + V_{OUT2} \times I_{OUT2} + V_{OUT3} \times I_{OUT3}$$

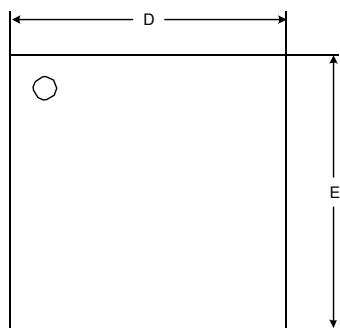
PCB Layout Recommendations

Layout is very critical for PMIC designs. For HM5301 PCB layout considerations, please refer to the following suggestions to get best performance.

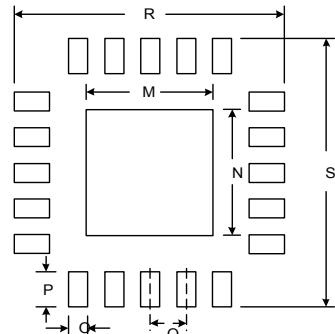
- ▶ It is suggested to use 4-layer PCB layout and place LX plane and output plane on the top layer, place VIN plane in the inner layer.
- ▶ The top layer SMD input and output capacitors ground plane should be connected to the internal ground layer and bottom ground plane individually by using vias.
- ▶ The AGND should be connected to inner ground layer directly by using via.
- ▶ High current path traces need to be widened.
- ▶ Place the input capacitors as close as possible to the VINx pin to reduce noise interference.
- ▶ Keep the feedback path (from V_{OUTX} to FB_x) away from the noise node (ex. LX_x). LX_x is a high current noise node. Complete the layout by using short and wide traces.
- ▶ The top layer exposed pad ground plane should be connected to the internal ground layer and bottom ground plane by using a number of vias to improve thermal performance.
- ▶ Place the input capacitors as close as possible to the VINx pin to reduce noise interference.

Package Information

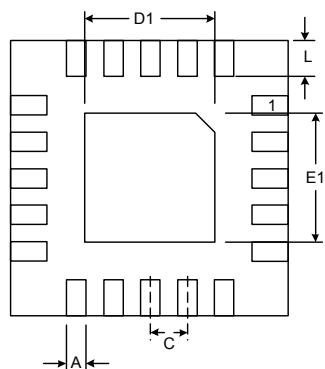
QFN 3mm x 3mm-20 Package



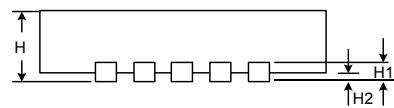
Top View



Recommended Layout Pattern



Bottom View



Side View

Unit: mm

Symbol	Dimension		Symbol	Dimension
	Min	Max		Typ
A	0.15	0.25	M	1.50
C	0.35	0.45	N	1.50
D	2.90	3.10	O	0.30
E	2.90	3.10	P	0.80
D1	1.55	1.75	Q	0.40
E1	1.55	1.75	R	3.90
L	0.35	0.45	S	3.90
H	0.70	0.80		
H1	0.18	0.25		
H2	0.00	0.05		