

#### **Features**

• Single-Supply Operation from +2.5V ~ +5.5V

• Rail-to-Rail Output

-3dB Bandwidth(G=+1): 350MHz (Typ.)

• Low Input Bias Current: 1pA (Typ.)

• Quiescent Current: 4.2mA/Amplifier (Typ.)

Operating Temperature: -40°C ~ +125°C

Small Package:

HM8091 Available in SOT23-5 and SC70-5 Packages
HM8092 Available in SOP-8 and MSOP-8 Packages
HM8094 Available in SOP-14 and TSSOP-14 Packages

HM8091N Available in SOT23-6 and SC70-6 Packages

HM8092N Available in MSOP-10 Packages

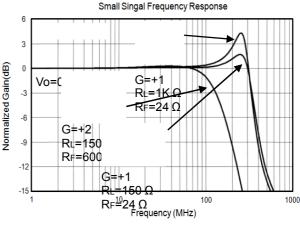
### **General Description**

The HM8091/1N(single), HM8092/2N(dual), HM8094(quad) are rail-to-rail output voltage feedback amplifiers offering ease of use and low cost. They have bandwidth and slew rate typically found in current feedback amplifiers. All have a wide input common-mode voltage range and output voltage swing, making them easy to use on single supplies as low as 2.5V. Despite being low cost, the HM8091 series provide excellent overall performance. They offer wide bandwidth to 350 MHz (G = +1) along with 0.1dB flatness out to 58 MHz (G = +2) and offer a typical low power of 4.2 mA/amplifier.

The HM8091 series is low distortion and fast settling make it ideal for buffering high speed A/D or D/A converters. The HM8091/2N has a power-down disable feature that reduces the supply current to  $75\mu$ A. These features make the HM8091/2N ideal for portable and battery-powered applications where size and power are critical. All are specified over the extended -40  $^{\circ}$ C to +125  $^{\circ}$ C temperature range.

### **Applications**

- Imaging
- Photodiode Preamp
- DVD/CD
- Filters
- Professional Video and Cameras
- Hand Sets
- Base Stations
- A-to-D Driver



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### **Pin Configuration**

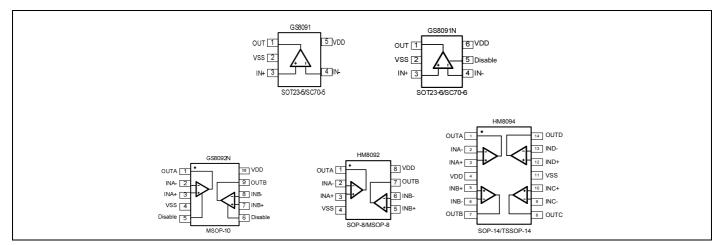


Figure 1. Pin Assignment Diagram

### **Absolute Maximum Ratings**

Condition	Min	Max			
Power Supply Voltage (V <sub>DD</sub> to Vss)	-0.5V	+7.5V			
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V <sub>DD</sub> +0.5V			
PDB Input Voltage	Vss-0.5V	+7V			
Operating Temperature Range	-40°C	+125°C			
Junction Temperature	+160	)°C			
Storage Temperature Range	-55°C	+150°C			
Lead Temperature (soldering, 10sec)	+260	)°C			
Package Thermal Resistance (T <sub>A</sub> =+25℃)					
SOP-8, θ <sub>JA</sub>	125°0	C/W			
MSOP-8, θ <sub>JA</sub>	216°0	C/W			
SOT23-5, θ <sub>JA</sub>	190°0	C/W			
SOT23-6, θ <sub>JA</sub>	190°0	C/W			
SC70-5, θ <sub>JA</sub>	333°C/W				
ESD Susceptibility	•				
НВМ	6K	6KV			
MM	400	400V			

**Note:** Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

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### **Package/Ordering Information**

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
HM8091	Single	HM8091-UR	SC70-5	Tape and Reel,3000	8091
	33	HM8091-MR	SOT23-5	Tape and Reel,3000	8091
HM8092	Dual	HM8092-SR	SOP-8	Tape and Reel,4000	GS8092
	2 5.5.	HM8092-MR	MSOP-8	Tape and Reel,3000	GS8092
HM8094	Quad	HM8094-TR	TSSOP-14	Tape and Reel,3000	GS8094
	4	HM8094-SR	SOP-14	Tape and Reel,2500	GS8094
HM8091N	Single With	HM8091N-UR	SC70-6	Tape and Reel,3000	8091N
	shutdown	HM8091N-MR	SOT23-6	Tape and Reel,3000	8091N
HM8092N	Dual With shutdown	HM8092N-MR	MSOP-10	Tape and Reel,2500	GS8092N

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### **Electrical Performance Characteristics**

 $(G=+2, R_F=600\Omega, R_G=600\Omega, and R_L=150\Omega)$  connected to  $V_S/2$ , unless otherwise noted. Typical values are at  $T_A=+25$ °C.)

		HM8091/8092/8094/8091N/8092N						
PARAMETER	CONDITIONS	TYP MIN/MAX OVER TEMPERATURE						
				0℃	-40℃to	-40℃		MIN/
		+25℃	+25℃	to70℃	85℃	to125℃	UNITS	MAX
DYNAMIC PERFORMANCE								
-3dB Small Signal Bandwidth	G = +1, Vo = 0.1V p-p, $R_F$ = 24 Ω, $R_L$ = 150Ω	335					MHz	TYP
	G = +1, Vo = 0.1V p-p, $R_F$ = 24 $\Omega$ , $R_L$ = 1k $\Omega$	330					MHz	TYP
	G = +2, Vo = 0.1V p-p, $R_L$ = 50 $\Omega$	79					MHz	TYP
	$G = +2$ , $Vo = 0.1V p-p$ , $R_L = 150\Omega$	130					MHz	TYP
	G = +2, Vo = 0.1V p-p, $R_L$ = 1k $\Omega$	165					MHz	TYP
	$G = +2$ , $Vo = 0.1V p-p$ , $R_L = 10k\Omega$	172					MHz	TYP
Gain-Bandwidth Product	$G = +10, R_L = 150\Omega$	180					MHz	TYP
	$G = +10$ , $R_L = 1k\Omega$	195					MHz	TYP
Bandwidth for 0.1dB Flatness	G = +2, Vo = 0.1V $_{p-p}$ , $R_L$ = 150 $\Omega$ , $R_F$ =600 $\Omega$	71					MHz	TYP
Slew Rate	G = +1, 2V Output Step	119/-232					V/ μ s	TYP
	G = +2, 2V Output Step	135/-180					V/ μ s	TYP
	G = +2, 4V Output Step	142/-206					V/ μ s	TYP
Rise-and-Fall Time	G = +2, Vo = $0.2V_{p-p}$ , 10% to 90%	3.5					ns	TYP
	$G = +2$ , $Vo = 2V_{p-p}$ , 10% to 90%	8.5					ns	TYP
Settling Time to 0.1%	G = +2, 2V Output Step	35					ns	TYP
Overload Recovery Time	V <sub>IN</sub> • G = +VS	14.5					ns	TYP
NOISE/DISTORTION PERFORMANCE								
Input Voltage Noise	f = 1MHz	4.3					nV/ Hz	TYP
Differential Gain Error (NTSC)	$G = +2, R_L = 150\Omega$	0.004					%	TYP
Differential Phase Error (NTSC)	$G = +2, R_L = 150\Omega$	0.08					degree	TYP
DC PERFORMANCE								
Input Offset Voltage (Vos)		±2	±8	±8.5	±9	±9.3	mV	MAX
Input Offset Voltage Drift		2					μ <b>V/</b> °C	TYP
Input Bias Current (I <sub>B</sub> )		1					PA	TYP
Input offset Current (I <sub>OS</sub> )		2					PA	TYP
Open-Loop Gain (A <sub>OL</sub> )	$V_{\rm O}$ = 0.3V to 4.7V, $R_{\rm L}$ = 150 $\Omega$	80 75		74	74	70	dB	MIN
	$V_{\rm O}$ = 0.2V to 4.8V, $R_{\rm L}$ = 1k $\Omega$	104	92	91	91	80	dB	MIN
INPUT CHARACTERISTICS								
Input Common-Mode Voltage Range ( $V_{\text{CM}}$ )		-0.2 to +3.8					V	TYP
Common-Mode Rejection Ratio (CMRR)	V <sub>CM</sub> = -0.1V to +3.5V	80	66	65	65	62	dB	MIN

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### **Electrical Performance Characteristics**

 $(G=+2, R_F=600\Omega, R_G=600\Omega, and R_L=150\Omega)$  connected to  $V_S/2$ , unless otherwise noted. Typical values are at  $T_A=+25^{\circ}C$ .)

		HM8091/8092/8094/8091N/8092N						
PARAMETER	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE			IRE		
				0℃	-40℃to	-40℃		MIN/
		+25℃	+25℃	to70℃	85℃	to125℃	UNITS	MAX
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail	$R_L = 150\Omega$	0.12					V	TYP
	$R_L = 1k\Omega$	0.03					٧	TYP
Output Current		120	100	98	93	87	mA	MIN
Closed-Loop Output Impedance	f<100kHz	0.045					Ω	TYP
POWER-DOWN DISABLE								
(HM8091/HM8092N only)								
Turn-On Time		108					ns	TYP
Turn-Off Time		60					ns	TYP
DISABLE Voltage-Off			0.8				V	MAX
DISABLE Voltage-On			2				V	MIN
POWER SUPPLY								
Operating Voltage Range			2.5	2.7	2.7	2.7	٧	MIN
			5.5	5.5	5.5	5.5	٧	MAX
Quiescent Current (per amplifier)		4.2	5.3	5.6	5.7	6.1	mA	MAX
Supply Current when Disabled per		75	120	130	132	137	μА	MAX
amplifier(HM8091/HM8092N only)								
Power Supply Rejection Ratio (PSRR)	$\Delta V_S = +2.7V \text{ to } +5.5V, V_{CM} = (-V_S) +0.5$	80	67	67	65	62	dB	MIN

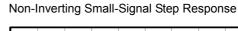
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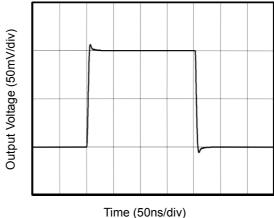


### **Typical Performance characteristics**

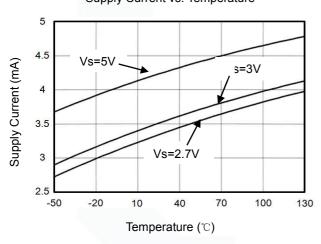
 $(Vs=+5V,G=+2,R_F=600\Omega,R_G=600\Omega,and\ R_L=150\Omega\ connected\ to\ Vs/2,\ unless\ otherwise\ noted.$  Typical values are at  $T_A=+25^{\circ}C.)$ Non-Inverting Large-Signal Step Response



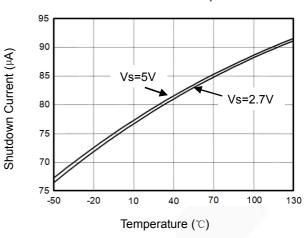




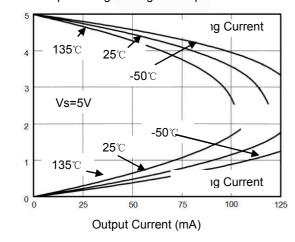
Supply Current vs. Temperature



Sutdown Current vs. Temperature

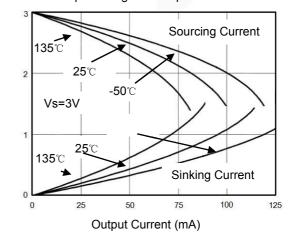


Output Voltage Swing vs. Output Current



Output Voltage (V)

Output Voltage vs. Output Current



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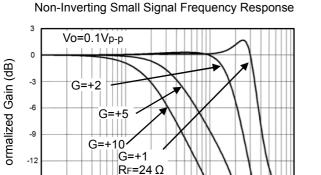
Output Voltage (V)



### **Typical Performance characteristics**

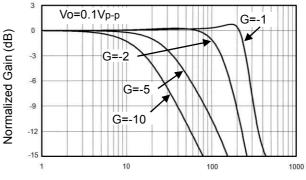
 $(Vs=+5V,G=+2,\,R_F=600\Omega,R_G=600\Omega,and\,R_L=150\Omega\,\,connected\,\,to\,\,Vs/2,\,unless\,\,otherwise\,\,noted.\,\,Typical\,\,values\,\,are\,\,at\,\,T_A=+25^{\circ}C.)$ 

1000

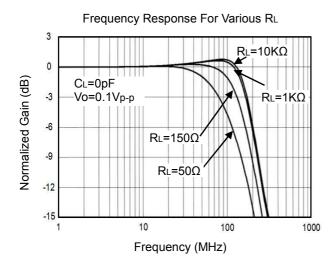




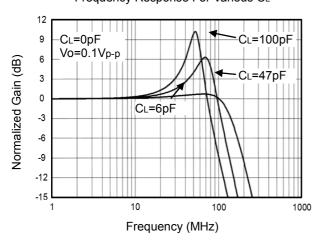




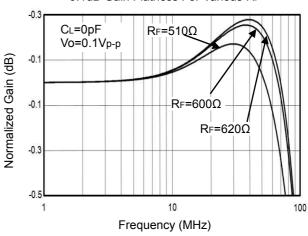
Frequency(MHz)



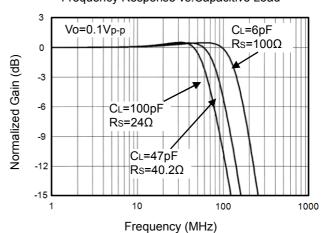
Frequency Response For Various CL



0.1dB Gain Flatness For Various RF



Frequency Response vs. Capacitive Load

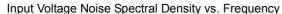


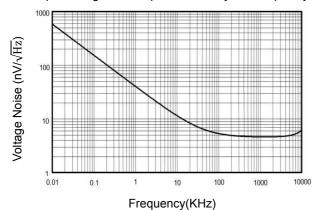
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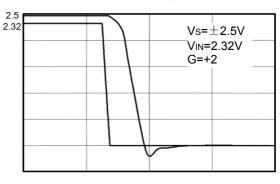
### **Typical Performance characteristics**

 $(Vs=+5V,G=+2,\,R_F=600\Omega,R_G=600\Omega,and\,\,R_L=150\Omega\,\,connected\,\,to\,\,Vs/2,\,unless\,\,otherwise\,\,noted.\,\,Typical\,\,values\,\,are\,\,at\,\,T_A=+25^{\circ}C.)$ 



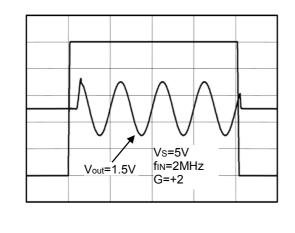


### Overload Recovery Time



Time(20ns/div)

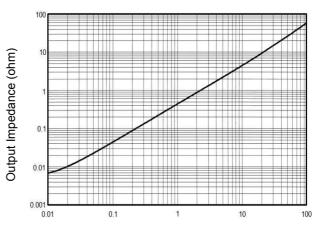
Large-Signal Disable/Enable Response



Output Voltage (1V/div)

Time (500n/div)

Closed-Loop Output Impedance vs Frequency



Frequency (MHz)

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### **Application Note**

### **Driving Capacitive Loads**

HM809X series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the HM809X series packages save space on printed circuit boards and enable the design of smaller electronic products.

### **Power Supply Bypassing and Board Layout**

HM809X series operates from a single 2.5V to 5.5V supply or dual  $\pm 1.25$ V to  $\pm 2.75$ V supplies. For best performance, a  $0.1\mu$ F ceramic capacitor should be placed close to the  $V_{DD}$  pin in single supply operation. For dual supply operation, both  $V_{DD}$  and  $V_{SS}$  supplies should be bypassed to ground with separate  $0.1\mu$ F ceramic capacitors.

### **Low Supply Current**

The low supply current (typical 4.2mA per channel) of HM809X series will help to maximize battery life. They are ideal for battery powered systems

#### **Operating Voltage**

HM809X series operate under wide input supply voltage (2.5V to 5.5V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime

### **Rail-to-Rail Output**

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of HM809X series can typically swing to less than 30mV from supply rail in light resistive loads (>1k $\Omega$ ), and 120mV of supply rail in moderate resistive loads (150 $\Omega$ ).

#### **Capacitive Load Tolerance**

The HM809X family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

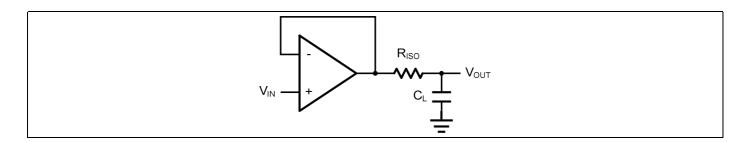


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the RISO resistor value, the more stable VOUT will be. However, if there is a resistive load RL in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

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The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the  $V_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

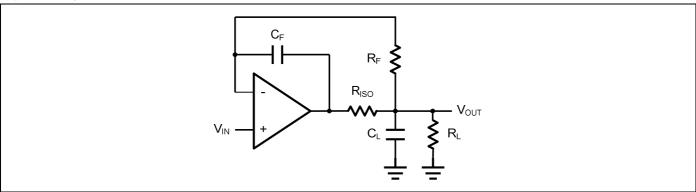


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy

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### **Typical Application Circuits**

### **Differential amplifier**

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using HM809X.

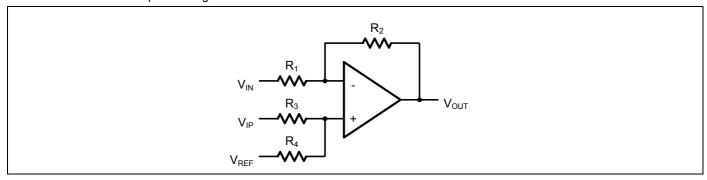


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R<sub>1</sub>=R<sub>3</sub> and R<sub>2</sub>=R<sub>4</sub>), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

#### **Low Pass Active Filter**

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C=1/(2\pi R_3C_1)$ .

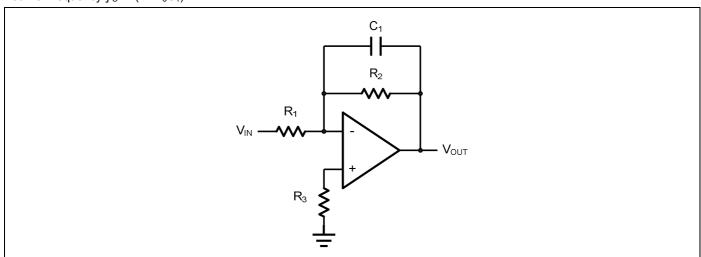


Figure 5. Low Pass Active Filter

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### **Driving Video**

The HM809X can be used in video applications like in Figure 6.

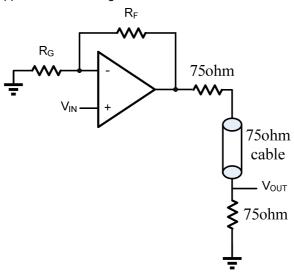


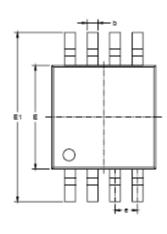
Figure 6. Typical video driving

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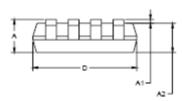


### **Package Information**

### MSOP8





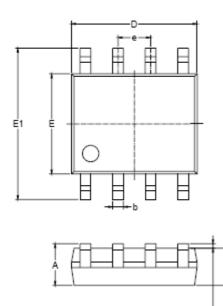


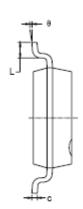
Symbol	Dimen In Milli		Dimensions In Inches		
•	MIN	MAX	MIN	MAX	
Α	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.008	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.650	0.650 BSC		BSC	
L	0.400	0.800	0.016	0.031	
θ	0°	6°	0°	6°	

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### SOP8



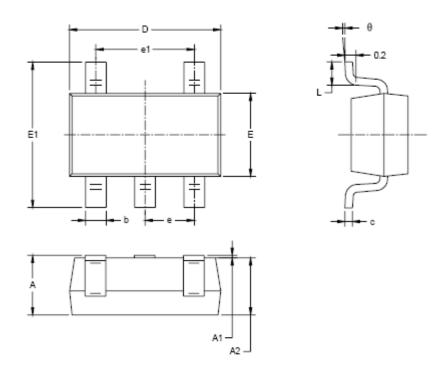


Symbol		nsions meters	Dimensions In Inches		
•	MIN	MAX	MIN	MAX	
Α	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
e	1.27	1.27 BSC		BSC	
L	0.400	1.270	0.016	0.050	
ө	0°	8°	0°	8°	

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### SOT23-5

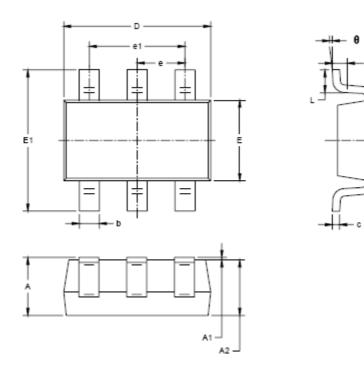


Symbol		nsions imeters	Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
A	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
E	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
е	0.950	BSC	0.037 BSC		
e1	1.900	1.900 BSC		BSC	
L	0.300	0.600	0.012	0.024	
ө	0°	8°	0°	8°	

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### SOT23-6

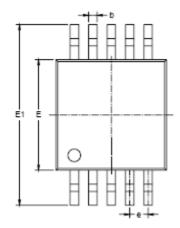


Symbol	Dimen In Milli		Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
Α	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
E	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
e	0.950	BSC	0.037 BSC		
e1	1.900	1.900 BSC		BSC	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	

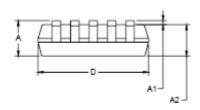
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### MSOP-10





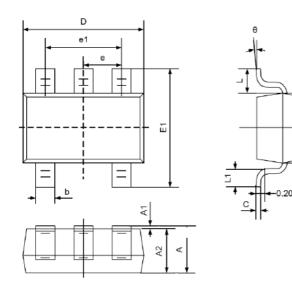


Symbol		nsions imeters	Dimensions In Inches		
•	MIN	MAX	MIN	MAX	
Α	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.006	
A2	0.750	0.950	0.030	0.037	
b	0.180	0.280	0.007	0.011	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.500	0.500 BSC		BSC	
L	0.400	0.800	0.016	0.031	
θ	0°	6°	0°	6°	

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### SC70-5

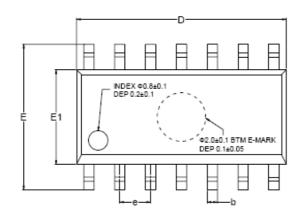


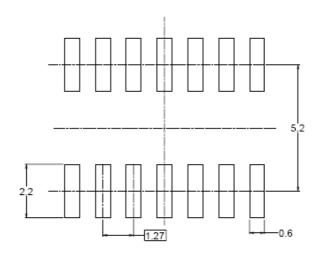
	Dimens	sions	Dimens	sions
Symbol	In Milli	In Millimeters		es
	Min	Max	Min	Max
Α	0.900	1.100	0.035	0.043
A1	0.000	0.100	0.000	0.004
A2	0.900	1.000	0.035	0.039
b	0.150	0.350	0.006	0.014
С	0.080	0.150	0.003	0.006
D	2.000	2.200	0.079	0.087
E	1.150	1.350	0.045	0.053
E1	2.150	2.450	0.085	0.096
е	0.650T	ΥP	0.026T	ΥP
e1	1.200	1.400	0.047	0.055
L	0.525REF		0.021REF	
L1	0.260	0.460	0.010	0.018
θ	0°	8°	0°	8°

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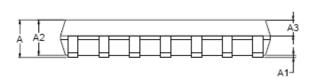


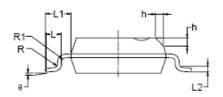
#### **SOP-14**





RECOMMENDED LAND PATTERN (Unit: mm)



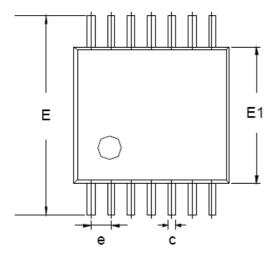


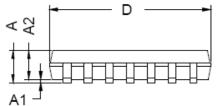
Cymah al	Symbol Dimensions In Millimeters		Dime	nsions In Ir	nches	
Symbol	MIN	MOD	MAX	MIN	MOD	MAX
Α	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.004		0.010
A2	1.25		1.65	0.049		0.065
A3	0.55		0.75	0.022		0.030
b	0.36		0.49	0.014		0.019
D	8.53		8.73	0.336		0.344
E	5.80		6.20	0.228		0.244
E1	3.80		4.00	0.150		0.157
е	1.27 BSC				0.050 BSC	
L	0.45		0.80	0.018		0.032
L1	1.04 REF				0.040 REF	
L2		0.25 BSC			0.01 BSC	
R	0.07			0.003		
R1	0.07			0.003		
h	0.30		0.50	0.012		0.020
θ	0°		8°	0°		8°

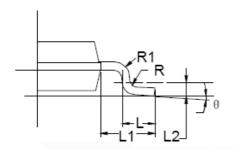
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### TSSOP-14







	Dimensions					
Symbol	In Millimeters					
Symbol	MIN	TYP	MAX			
Α	-	-	1.20			
A1	0.05	-	0.15			
A2	0.90	1.00	1.05			
b	0.20	-	0.28			
С	0.10	-	0.19			
D	4.86	4.96	5.06			
E	6.20	6.40	6.60			
E1	4.30	4.40	4.50			
е		0.65 BSC				
L	0.45	0.60	0.75			
L1	1.00 REF					
L2	0.25 BSC					
R	0.09	-	-			
θ	0°	-	8°			

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